

In the Name of GOD

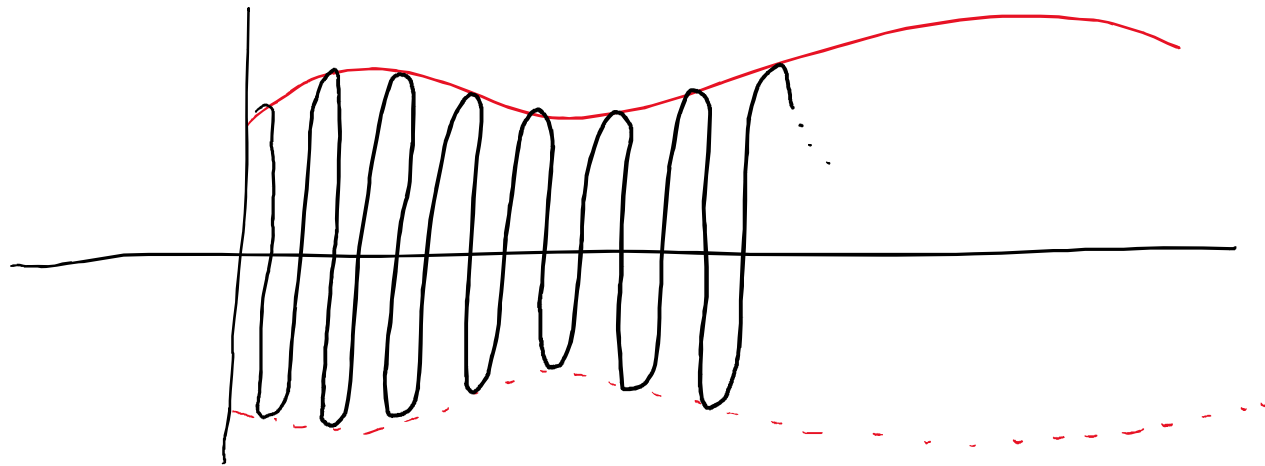
Energy of bandpass signal based on the Energy of equ. baseband signal

$$\underbrace{s(t)}_{\text{bandpass signal}} = \operatorname{Re} \left\{ \underbrace{s_e(t)}_{\text{equ. baseband (low pass) signal}} e^{j2\pi f_c t} \right\}$$

$$\underbrace{E_s}_{\text{Energy of } s(t)} = \frac{1}{2} \underbrace{E_{s_e}}_{\text{Energy of } s_e(t)}$$

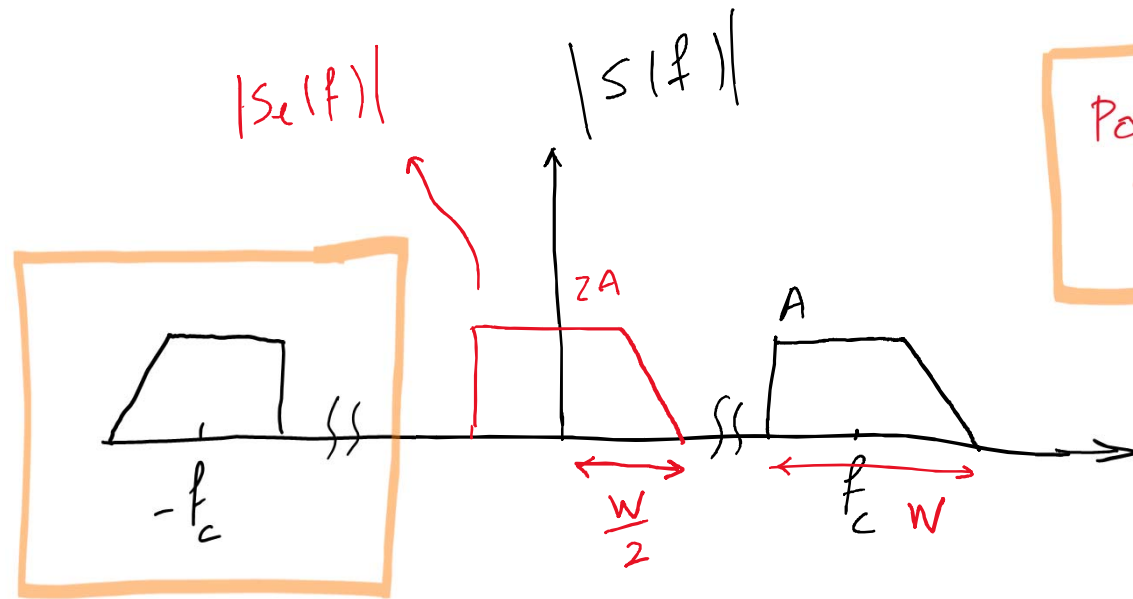
$$\int a^2(t) \cos(4\pi f_c t + 2\theta(t)) dt = 0$$

because $W \ll \frac{1}{T_c}$ that means time variation of $a(t)$ is so smaller than that of $\cos(4\pi f_c t + 2\theta(t))$. So the above integration is equal to zero.



$a^2(t)$:
 $a(t)$: envelope of $s_e(t)$

$a(t)$ is approximately
 Constant in Comparison to
 $\cos(4\pi f_c t + 2\theta(t))$



Power of $s(t)$ $\leftarrow P_s = P_{s_e}$
 Power of $s_e(t)$

$$\epsilon_s = \frac{1}{2} \epsilon_{s_e}$$

BW of $s_e(t)$ is $\frac{1}{2}$ of
 that of $s(t)$

now, we want to find an expression for spectrum of bandpass signal $s(t)$ based on the spectrum of eqn. baseband signal. As you know, we have

$$s(t) = \text{Re} \left\{ s_b(t) e^{j2\pi f_c t} \right\}$$

$$\xRightarrow{\text{FT}\{\cdot\}} \text{FT}\{s(t)\} = \text{FT} \left\{ \text{Re} \left\{ s_b(t) e^{j2\pi f_c t} \right\} \right\}$$

$$\Rightarrow S(f) = \text{FT} \left\{ \text{Re} \left\{ s_b(t) e^{j2\pi f_c t} \right\} \right\}$$

We know for any complex parameter z , we have

$$\underline{\operatorname{Re}\{z\} = \frac{1}{2}\{z + z^*\}}$$

$$\Rightarrow S(f) = FT \left\{ \underbrace{\operatorname{Re}\{s_e(t) e^{j2\pi f_c t}\}}_{\frac{1}{2}\{s_e(t) e^{j2\pi f_c t} + s_e^*(t) e^{-j2\pi f_c t}\}} \right\}$$

$$\Rightarrow S(f) = \frac{1}{2} FT \left\{ s_e(t) e^{j2\pi f_c t} + s_e^*(t) e^{-j2\pi f_c t} \right\}$$

$$S(f) = \frac{1}{2} [s_e(f - f_c) + s_e^*(-f - f_c)]$$

FT characteristics

$$\Rightarrow S(f) = \frac{1}{2} [S_c(f - f_c) + \underline{S_c^*(-f - f_c)}]$$

$$1) \quad s(t) = \operatorname{Re} \{ s_c(t) e^{j2\pi f_c t} \}$$

$$2) \quad s(t) = x(t) \cos 2\pi f_c t - y(t) \sin 2\pi f_c t$$

$$3) \quad s(t) = a(t) \cos(2\pi f_c t + \theta(t))$$

$$s_c(t) = x(t) + j y(t) \quad , \quad a(t) = \sqrt{x^2(t) + y^2(t)} \quad , \quad \theta(t) = \tan^{-1} \frac{y(t)}{x(t)}$$

Simulation Assignment # 1

We want to write a code in Python (or MATLAB) to plot the power spectrum of a signal $s(t)$.

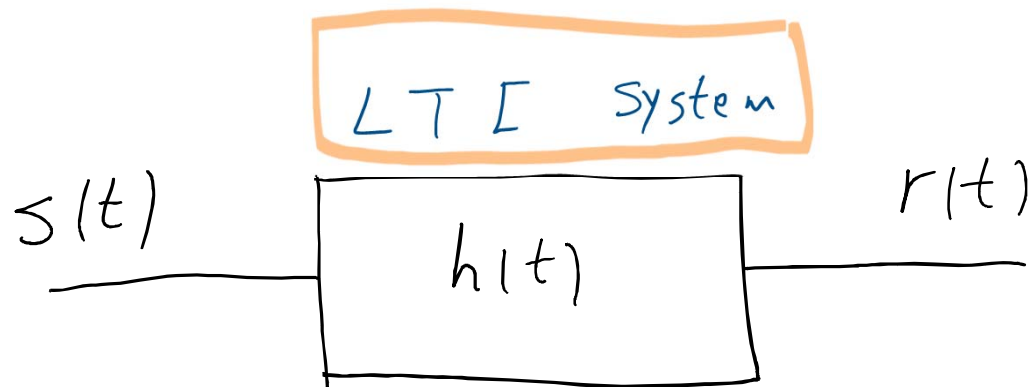
In the first step, suppose that $s(t)$ is a deterministic signal.

And in the second step modify your code so that it

can plot the PSD of an stochastic signal $s(t)$.

* How to represent a bandpass system, based on equ. baseband system

In this part we want to talk about Linear Time Invariant bandpass systems and their equ. baseband systems.



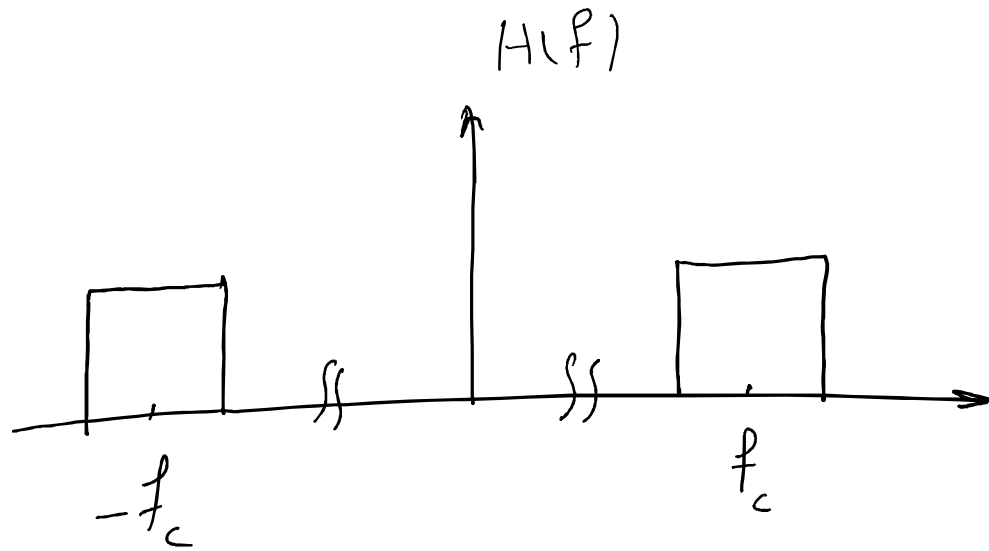
$$r(t) = s(t) * h(t)$$

$$\underline{R(f) = S(f) H(f)} \quad (*)$$

The output of any band-pass is obviously a band-pass signal. So $r(t)$ is a bandpass signal.

As you know any LTI system is known by its Impulse

Response $h(t)$



$$\underbrace{H(f)}_{\substack{\uparrow \\ \text{Transfer function of} \\ \text{the system}}} = \text{FT} \{ h(t) \}$$

Also, in this part we talk about band-pass systems that their transfer functions is located in high frequencies and have Hermitian symmetry. (because Impulse Response $h(t)$ is a real function)

So, we want to find an expression for $r(f)$ based on equ. lowpass signal and the equ. lowpass system.

We start from the equ. low-pass system.

$$H_e(f - f_c) = \begin{cases} H(f) & f > 0 \\ 0 & f < 0 \end{cases}$$

①

Hermitian Symmetry

$$\xrightarrow{\text{Hermitian Symmetry}} H^*(-f) = H(f) \quad (*)$$

$$H_e^*(-f - f_c) = \begin{cases} 0 & f > 0 \\ \underbrace{H^*(-f)}_{H(f)} & f < 0 \end{cases}$$

$f > 0$
 $f < 0$

②

$$\textcircled{1}, \textcircled{2} \implies H(f) = H_p(f - f_c) + H_p^*(-f - f_c)$$

$$\stackrel{\text{FT}}{\implies} h(t) = 2 \operatorname{Re} \left\{ h_p(t) e^{j2\pi f_c t} \right\}$$

Now, we want to find an expression for $r(t)$ and its equ. lowpass signal and equ. lowpass system.

$$\left. \begin{aligned} r(t) &= s(t) * h(t) \\ R(f) &= S(f) H(f) \end{aligned} \right\} \begin{array}{l} \text{all are} \\ \text{band-pass signals} \\ \text{and system} \end{array}$$

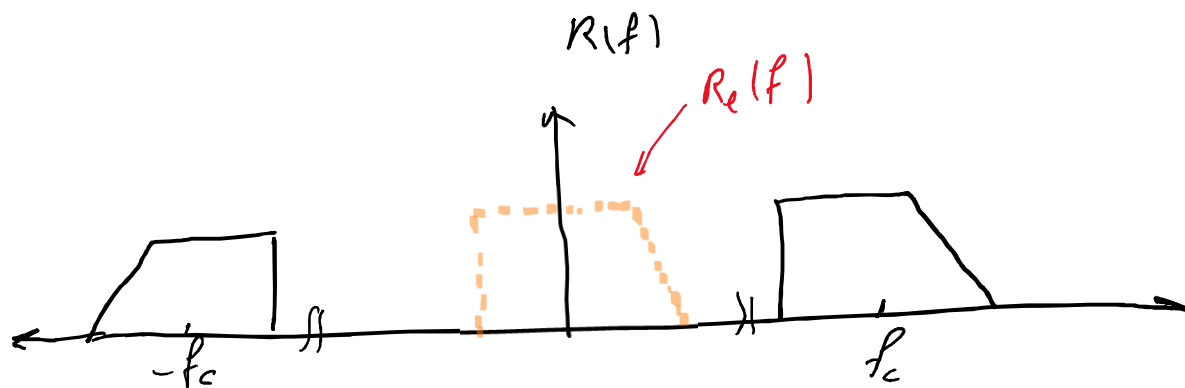
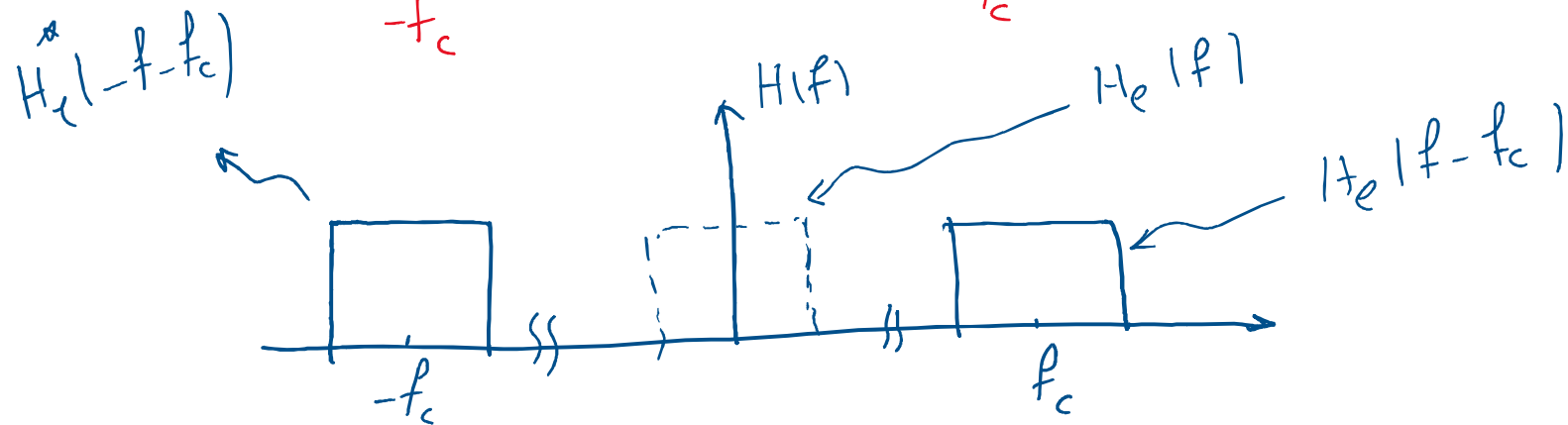
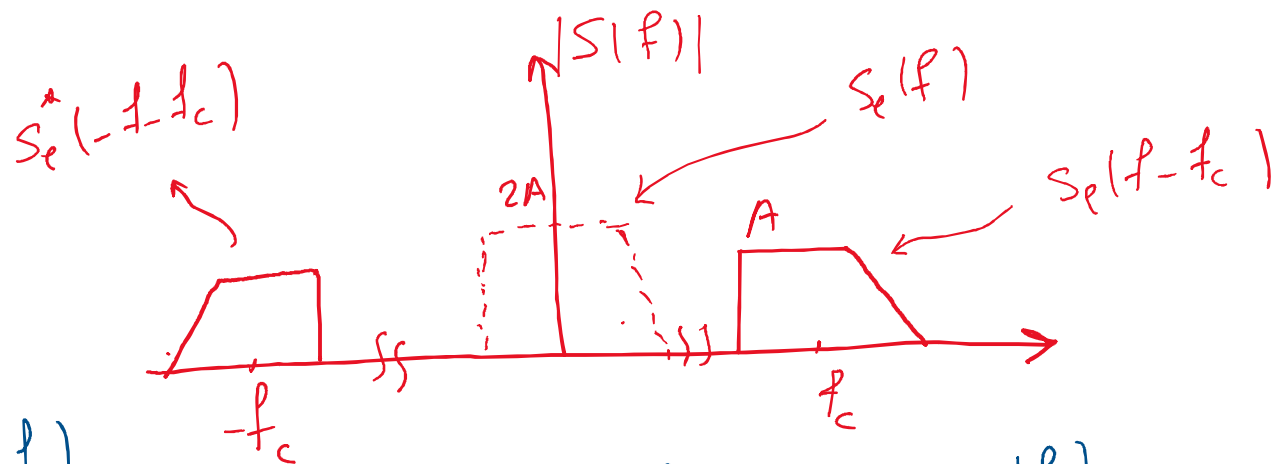
$$R(f) = S(f) H(f)$$

$$= \frac{1}{2} \left[S_e(f-f_c) + S_e^*(-f-f_c) \right] \left[H_e(f-f_c) + H_e^*(-f-f_c) \right]$$

$$= \frac{1}{2} \left[S_e(f-f_c) H_e(f-f_c) + S_e^*(-f-f_c) H_e^*(-f-f_c) \right]$$

$$+ \frac{1}{2} \left[\underbrace{S_e(f-f_c) H_e^*(-f-f_c)}_0 + \underbrace{S_e^*(-f-f_c) H_e(f-f_c)}_0 \right]$$

$$\Rightarrow R(f) = \frac{1}{2} \left[S_e(f-f_c) H_e(f-f_c) + S_e^*(-f-f_c) H_e^*(-f-f_c) \right]$$



$$R(f) = S(f) H(f)$$

$$R(f) = \frac{1}{2} [S_e(f-f_c) H_e(f-f_c) + S_e^*(-f-f_c) H_e^*(-f-f_c)]$$

$$R_e(f) = S_e(f) H_e(f)$$

$\Rightarrow$

$$R(f) = \frac{1}{2} [R_e(f-f_c) + \hat{R}_e(-f-f_c)]$$

$$\Rightarrow r(t) = \text{Re} \{ r_e(t) e^{j2\pi f_c t} \}$$

$$r_e(t) = s_e(t) * h_e(t)$$

So, we can conclude that the performance of a band-pass commu. system, is equal to that of baseband equivalent system.

So, we can do our mathematical analysis of a band-pass system, in the equ. base-band form.

Because the math. analysis in baseband freq. is so simpler than that of in band-pass freq. we know that no information will be missed by this operation.